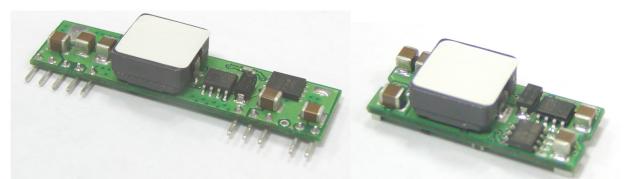


Application Note Ver12 July 2005

## **NON-ISOLATED DC-DC Converter**

SIP16-12S05A, SMT16-12S05A 9.0-14Vin, 0.75- 5.0Vout, 16A

> APPLICATION NOTE Ver12



SIP16-12S05A

SMT16-12S05A

DOCUMENT HISTORY				
VERSION	DATE	SUMMARY of CHANGE		
Ver10	03/15/04	New Release		
Ver11	03/15/04	Change Missing Data		
Ver12	07/15/05	Add Technical Specifications		



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### 1. Introduction

This application note describes the features and functions of Cincon's SIP/SMT16-12 series of Non Isolated DC-DC Converters. These are highly efficient, reliable and compact, high power density, single output DC/DC converters. These "Point of Load" modules serve the needs specifically of the fixed and mobile telecommunications and computing market, employing economical distributed Power Architectures. The SIP/SMT16-12 series provide precisely regulated output voltage range from 0.75V to 5.0Vdc over a wide range of input voltage (Vi=9.0 – 14Vdc) and can operate over an ambient temperature range of  $-40^{\circ}$ C to +85°C. Ultra-high efficiency operation is achieved through the use of synchronous rectification and drive control techniques.

The modules are fully protected against short circuit and overtemperature conditions. Cincon's world class automated manufacturing methods, together with an extensive testing and qualification program, ensure that all SIP/SMT16-12 series converters are extremely reliable.

### 2. 16A SIP/SMT Converter Features

- High efficiency topology, typically 94% at 5.0Vdc
- Industry standard footprint
- Wide ambient temperature range, -40C to +85C
- Cost efficient open frame design
- Programmable output voltage via external resistor from 0.75 to 5.0Vdc
- No minimum load requirement (Stable at all loads)
- Remote ON/OFF
- Remote sense compensation
- Fixed switching frequency
- Continuous short-circuit protection and over current protection
- Over-temperature protection (OTP)
- Monotonic Startup with pre-bias at the output.
- UL/IEC/EN60950 Certified.

### 3. General Description

#### **3.1 Electrical Description**

A block diagram of the SIP/SMT16-12 Series converter is shown in Figure 1. Extremely high efficiency power conversion is achieved through the use of synchronous rectification and drive techniques. Essentially, the powerful SIP/SMT16-12 series topology is based on a non-isolated synchronous buck converter. The control loop is optimized for unconditional stability, fast transient response and a very tight line and load regulation. In a typical pre-bias application the SIP/SMT16-12 series converters do not draw any reverse current at start-up. The output voltage can be adjusted from 0.75 to 5.0vdc, using the TRIM pin with a external resistor. The converter can be shut down via a remote ON/OFF input that is referenced to ground. This input is compatible with popular logic devices; a 'positive' logic input is supplied as standard. Positive logic implies that the converter is enabled if the remote ON/OFF input is high (or floating), and disabled if it is low. The converter is also protected against over-temperature conditions. If the converter is overloaded or the ambient temperature gets too high, the converter will shut down to protect the unit.

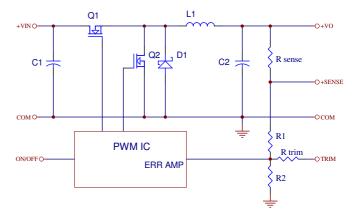


Figure 1. Electrical Block Diagram

#### 3.2 Thermal Packaging and Physical Design.

The SIP/SMT16-12 series uses a multi-layer FR4 PCB construction. All surface mount power components are placed on one side of the PCB, and all low-power control components are placed on the other side. Thus, the Heat dissipation of the power components is optimized, ensuring that control components are not thermally stressed. The converter is an open-frame product and has no case or case pin. The open-frame design has several advantages over encapsulated closed devices. Among these advantages are:

- Efficient Thermal Management: the heat is removed from the heat generating components without heating more sensitive, small signal control components.
- Environmental: Lead free open-frame converters are more easily re-cycled.
- Cost Efficient: No encapsulation. Cost efficient open-frame construction.
- **Reliable:** Efficient cooling provided by open frame construction offers high reliability and easy diagnostics.



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### 4. Technical Specifications

(All specifications are typical at nominal input, full load at 25°C unless otherwise noted.)

NOTES and CONDITIONS	Device	Min.	Typical	Max.	Units
	ALL	0		15	Vdc
	ALL	-40		+85	°C
	ALL	-55		+125	°C
· · ·					
	ALL	9.0	12	14	Vdc
					Vdc
					Vdc Vdc
Vin 0 to 14)/do. lo lo mov			0.3	11	A
	,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,			11	A
Vo=0.75V Vo=1.2V Vo=1.5V Vo=2.0V Vo=2.0V Vo=2.5V Vo=3.3V Vo=5.0V	ALL		40 50 50 60 60 65 75 75		mA
Shutdown input idle current	ALL			10	mA
	ALL			0.1	A <sup>2</sup> s
P-P thru 1uH inductor, 5Hz to 20MHz	ALL		300		mA
Vin=Nominal Vin , lo=Io.max, Tc=25 $^\circ\!\!\mathbb{C}$	ALL	-1.5%	Vo,set	+1.5%	Vdc
Selected by an external resistor	ALL	0.75		5.0	Vdc
lo=lo.min to lo.max	ALL	-0.5		+0.5	%
Vin=low line to high line	ALL	-0.2		+0.2	%
Ta=-40°C to 85°C	ALL	-0.03		+0.03	<b>%/</b> °C
5Hz to 20MHz bandwidth					
Full Load, 1uF ceramic and 10uF tantalum	ALL			75	mV
Full Load, 1uF ceramic and 10uF tantalum	ALL			30	mV
Low ESR	ALL			8000	uF
	ALL	0		16	Α
Output Voltage =90% Nominal Output Voltage	ALL	18	23	28	Α
	ALL		1		
	Vin=0 to 14Vdc , lo=lo,max.     Vin=0 to 14Vdc , lo=lo,max.     Vo=0.75V     Vo=1.2V     Vo=1.5V     Vo=2.5V     Vo=2.5V     Vo=3.3V     Vo=5.0V     Shutdown input idle current     P-P thru 1uH inductor, 5Hz to 20MHz     Vin=Nominal Vin , lo=lo.max, Tc=25°C     Selected by an external resistor     Io=lo.min to lo.max     Vin=low line to high line     Ta=-40°C to 85°C     5Hz to 20MHz bandwidth     Full Load, 1uF ceramic and 10uF tantalum     Full Load, 1uF ceramic and 10uF tantalum     Low ESR	ALL     Vin=0 to 14Vdc , lo=lo,max.     Vo=0.75V     Vo=1.5V     Vo=1.5V     Vo=1.5V     Vo=1.8V     Vo=2.0V     Vo=5.0V     Shutdown input idle current     ALL     P-P thru 1uH inductor, 5Hz to 20MH	ALL   0     ALL   -40     ALL   -55     ALL   -55     ALL   -55     ALL   9.0     ALL   -55     ALL   9.0     ALL   -55     ALL   9.0     ALL   -10     ALL   -10     Vin=0 to 14Vdc , lo=lo,max.   ALL     Vo=0.75V   Vo=1.5V     Vo=1.5V   Vo=2.5V     Vo=2.5V   Vo=3.3V     Vo=2.5V   Vo=3.3V     Vo=5.0V   ALL     Shutdown input idle current   ALL     Vin=Nominal Vin , lo=lo.max, Tc=25°C   ALL     Vin=Nominal Vin , lo=lo.max, Tc=25°C   ALL   -1.5%     Selected by an external resistor   ALL   0.75     Vin=Nominal Vin , lo=lo.max   ALL   -0.5     Vin=Nominal Vin , lo=lo.max   ALL   0.75     Io=lo.min to lo.max   ALL   -0.5     Vin=low line to high line   ALL   -0.2     Ta=-40°C to 85°C   ALL   -0.03     5Hz to 20MHz bandwidth   -10.2 <td>ALL     0       ALL     -40       ALL     -40       ALL     -40       ALL     -55       ALL     -55       ALL     9.0     12       ALL     9.0     12       ALL     7.7       ALL     0.3       Vin=0 to 14Vdc , lo=lo,max.     ALL       Vo=0.75V     40       Vo=1.2V     50       Vo=1.2V     50       Vo=1.2V     50       Vo=2.0V     60       Vo=2.5V     50       Vo=3.3V     50       Vo=5.0V     75       Shutdown input idle current     ALL       P-P thru 1uH inductor, 5Hz to 20MHz     ALL       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -1.5%       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -0.5       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -0.5       Vin=low line to high line     ALL     -0.2       Ta=-40°C to 85°C     ALL     -0.03       SHz to 20MHz bandwidth     -0.2<td>ALL     0     15       ALL     -40     +85       ALL     -55     +125       ALL     -55     +125       ALL     -55     +125       ALL     -55     +125       ALL     -7.7     -       ALL     0.12     14       ALL     -7.7     -       ALL     0.3     -       Vin=0 to 14Vdc , lo=lo,max.     ALL     0.3       Vin=0 to 14Vdc , lo=lo,max.     ALL     11       Vo=0.75V     40     50       Vo=1.2V     50     50       Vo=1.2V     50     50       Vo=2.0V     40     60       Vo=2.0V     60     60       Vo=2.0V     60     57       Shutdown input idle current     ALL     10       ALL     0.1     -       P-P thru 1uH inductor, 5Hz to 20MHz     ALL     300       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -1.5%       Vin=Nominal vin , lo=lo.max, Tc=25°C     ALL     <td< td=""></td<></td></td>	ALL     0       ALL     -40       ALL     -40       ALL     -40       ALL     -55       ALL     -55       ALL     9.0     12       ALL     9.0     12       ALL     7.7       ALL     0.3       Vin=0 to 14Vdc , lo=lo,max.     ALL       Vo=0.75V     40       Vo=1.2V     50       Vo=1.2V     50       Vo=1.2V     50       Vo=2.0V     60       Vo=2.5V     50       Vo=3.3V     50       Vo=5.0V     75       Shutdown input idle current     ALL       P-P thru 1uH inductor, 5Hz to 20MHz     ALL       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -1.5%       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -0.5       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -0.5       Vin=low line to high line     ALL     -0.2       Ta=-40°C to 85°C     ALL     -0.03       SHz to 20MHz bandwidth     -0.2 <td>ALL     0     15       ALL     -40     +85       ALL     -55     +125       ALL     -55     +125       ALL     -55     +125       ALL     -55     +125       ALL     -7.7     -       ALL     0.12     14       ALL     -7.7     -       ALL     0.3     -       Vin=0 to 14Vdc , lo=lo,max.     ALL     0.3       Vin=0 to 14Vdc , lo=lo,max.     ALL     11       Vo=0.75V     40     50       Vo=1.2V     50     50       Vo=1.2V     50     50       Vo=2.0V     40     60       Vo=2.0V     60     60       Vo=2.0V     60     57       Shutdown input idle current     ALL     10       ALL     0.1     -       P-P thru 1uH inductor, 5Hz to 20MHz     ALL     300       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -1.5%       Vin=Nominal vin , lo=lo.max, Tc=25°C     ALL     <td< td=""></td<></td>	ALL     0     15       ALL     -40     +85       ALL     -55     +125       ALL     -55     +125       ALL     -55     +125       ALL     -55     +125       ALL     -7.7     -       ALL     0.12     14       ALL     -7.7     -       ALL     0.3     -       Vin=0 to 14Vdc , lo=lo,max.     ALL     0.3       Vin=0 to 14Vdc , lo=lo,max.     ALL     11       Vo=0.75V     40     50       Vo=1.2V     50     50       Vo=1.2V     50     50       Vo=2.0V     40     60       Vo=2.0V     60     60       Vo=2.0V     60     57       Shutdown input idle current     ALL     10       ALL     0.1     -       P-P thru 1uH inductor, 5Hz to 20MHz     ALL     300       Vin=Nominal Vin , lo=lo.max, Tc=25°C     ALL     -1.5%       Vin=Nominal vin , lo=lo.max, Tc=25°C     ALL <td< td=""></td<>



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PARAMETER	NOTES and CONDITIONS	Device	Min.	Typical	Max.	Units
DYNAMIC CHARACTERISTICS		·				
Output Voltage Transient Response						
Error Brand	50% Step Load Change, di/dt=0.1A/us	ALL			200	mV
Setting Time (within 1% Vout nominal)	50% Step Load Change, di/dt=0.1A/us	ALL			200	us
EFFICIENCY						
100% Load	Vo=0.75V Vo=1.2V Vo=1.5V Vo=1.8V Vo=2.0V Vo=2.5V Vo=3.3V Vo=5.0V	ALL		77 83 86 88 89 90 92 94		%
ISOLATION CHARACTERISTICS						
Input to Output	Non-isolation	ALL	0			Vdc
FEATURE CHARACTERISTICS						
Switching Frequency		ALL		300		KHz
ON/OFF Control, Positive Logic Remote On/Off Logic Low (Module Off) Logic High (Module On)	or Open Circuit	SIP16-12S05A SMT16-12S05A	0	Vin	0.4	Vdc Vdc
ON/OFF Control, Negative Logic Remote On/Off Logic Low (Module On) Logic High (Module Off)	or Open Circuit	SIP16-12S05AN SMT16-12S05AN	0 2.8		0.4 Vin	Vdc Vdc
ON/OFF Current (for both remote on/off logic)	Ion/off at Von/off=0.0V	ALL			1	mA
Leakage Current (for both remote on/off logic)	Logic High, Von/off=14V	ALL			1	mA
Turn-On Delay and Rise Time Turn-On Delay Time, From On/Off Control	Von/off to 10%Vo.set	ALL		3.5		ms
Turn-On Delay Time, From Input	Vin.min. to 10%Vo.set	ALL		3.5		ms
Output Voltage Rise Time	10%Vo.set to 90%Vo.set	ALL		3.5		ms
Over Temperature Protection		ALL		130		°C
GENERAL SPECIFICATIONS						
MTBF	lo=100%of lo.max;Ta=25 °C per MIL-HDBK- 217F	ALL		0.98		M hours
Weight		ALL		10		grams
Dimensions						
SIP packages	2x0.512x0.327 inches (50.8x13.0x8.3 mm)					
SMT packages	1.3x0.53x0.366 inches(33.0x13.46x9.3 mm)					



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### 5. Main Features and Functions

#### 5.1 Operating Temperature Range

Cincon's SIP/SMT16-12 series converters highly efficient converter design has resulted in its ability to operate over a wide ambient temperature environment ( -40  $^{\circ}$ C to 85  $^{\circ}$ C). Due consideration must be given to the de-rating curves when ascertaining maximum power that can be drawn from the converter. The maximum power drawn is influenced by a number of factors, such as:

- Input voltage range.
- Output load current.
- Air velocity (forced or natural convection).
- Mounting orientation of converter PCB with respect to the Airflow.
- Motherboard PCB design, especially ground and power planes. These can be effective heat sinks for the converter.

#### 5.2 Over-Temperature Protection (OTP)

The SIP/SMT16-12 Series converters are equipped with non-latching over-temperature protection. A temperature sensor monitors the temperature of the hot spot (typically, top switch). If the temperature exceeds a threshold of 130°C (typical) the converter will shut down, disabling the output. When the temperature has decreased the converter will automatically restart.

The over-temperature condition can be induced by a variety of reasons such as external overload condition or a system fan failure.

#### 5.3 Output Voltage Adjustment

Section 7.8 describes in detail as to how to trim the output voltage with respect to its set point. The output voltage on all models is trimmable in the range 0.75 - 5.0Vdc.

#### 5.4 Safe Operating Area (SOA)

Figure 2 provides a graphical representation of the Safe Operating Area (SOA) of the converter. This representation assumes ambient operating conditions such as airflow are met as per thermal guidelines provided in Sections 7.2 and 7.3.

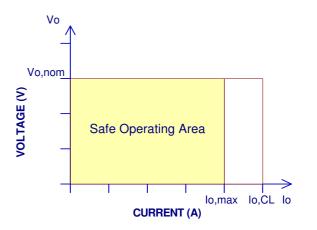


Figure 2. Maximum Output Current Safe Operating Area

#### **5.5 Over Current Protection**

All different voltage models have a full continuous short-circuit protection. The unit will auto recover once the short circuit is removed. To provide protection in a fault condition, the unit is equipped with internal over-current protection. The unit operates normally once the fault condition is removed. The power module will supply up to 150% of rated current. In the event of an over current converter will go into a hiccup mode protection.

#### 5.6 Remote ON/OFF

The remote ON/OFF input feature of the converter allows external circuitry to turn the converter ON or OFF. Active-high remote ON/OFF is available as standard. The SIP/SMT16-12S05A are turned on if the remote ON/OFF pin is high(=Vin), or left open. Setting the pin low(<0.4Vdc) will turn the converter 'Off'. The signal level of the remote on/off input is defined with respect to ground. If not using the remote on/off pin, leave the pin open (module will be on). The part number suffix "N" is Negative remote ON/OFF version. The unit is guaranteed OFF over the full temperature range if this voltage level exceeds 2.8Vdc. The converters are turned on If the on/off pin input is low (<0.4Vdc) or left open. The recommended SIP/SMT remote on/off drive circuit as shown as figure 3, 4.



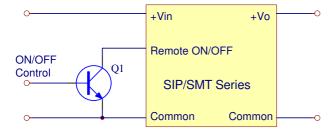


Figure 3. Positive Remote ON/OFF Input Drive Circuit

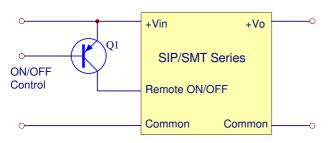


Figure 4. Negative Remote ON/OFF Input Drive Circuit

#### 5.7 UVLO (Under-Voltage Lockout)

The voltage on the Vcc pin determines the start of the operation of the Converter. When the input Vcc rises and exceeds about 8.0V the converter initiates a soft start. The UVLO function in the converter has a hysteresis (about 300mV) built in to provide noise immunity at start-up.

### 6. Safety

#### 6.1 Input Fusing and Safety Considerations.

<u>Agency Approvals:</u> The power Supply shall be submitted to and receive formal approval from the following test agencies.

1. The power supply shall be approved by a nationally recognized testing laboratory to UL/CSA 60950 3<sup>rd</sup> Edition (North America) and EN60950 (International)

2. CB Certificate from an internationally recognized test house in accordance with EN 60950.

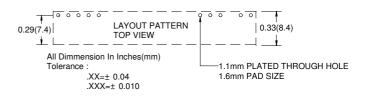
The SIP/SMT16-12 series converters do not have an internal fuse. However, to achieve maximum safety and system protection, always use an input line fuse. The safety agencies require a time-delay fuse with a maximum rating of 20A.

#### 7. Applications

#### 7.1 Layout Design Challenges.

In optimizing thermal design the PCB is utilized as a heat sink. Also some heat is transferred from the SIP/SMT module to the main board through connecting pins. The system designer or the end user must ensure that other components and metal in the vicinity of the SIP/SMT16-12 series meet the spacing requirements to which the system is approved.

Low resistance and low inductance PCB layout traces are the norm and should be used where possible. Due consideration must also be given to proper low impedance tracks between power module, input and output grounds. The recommended SIP/SMT footprint as shown as figure 5, 6.





Recommended Pad Layout

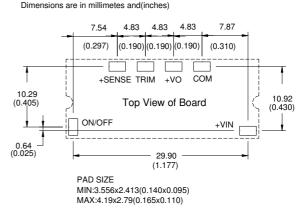


Figure 6. Recommended SMT Footprint

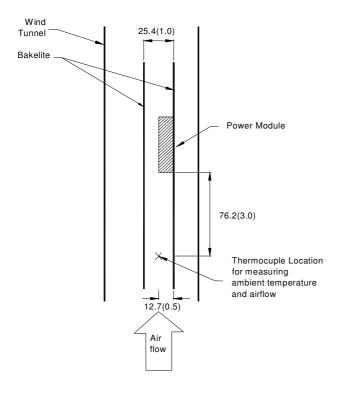


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#### 7.2 Convection Requirements for Cooling

To predict the approximate cooling needed for the module, refer to the Power De-rating curves in Figures 10 to 13. These de-rating curves are approximations of the ambient temperatures and airflows required to keep the power module temperature below its maximum rating. Once the module is assembled in the actual system, the module's temperature should be checked as shown in Figure 7 to ensure it does not exceed 120 °C.

Proper cooling can be verified by measuring the power module's temperature at Q1-pin 6 as shown in Figure 8,9.



Note : Dimensions are in millimeters and (inches)

Figure 7. Thermal Test Setup

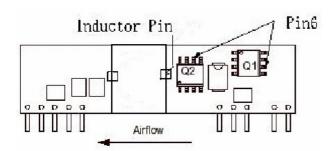


Figure 8. Temperature Measurement Location for SIP

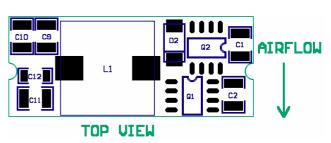


Figure 9. Temperature Measurement Location for SMT

#### 7.3 Thermal Considerations

The power module operates in a variety of thermal environments; however, sufficient cooling should be provided to help ensure reliable operation of the unit. Heat is removed by conduction, convection, and radiation to the surrounding environment. The thermal data presented is based on measurements taken in a set-up as shown in Figure7. Figures 10 to 13 represent the test data. Note that the airflow is parallel to the long axis of the module as shown in Figure7 for the SIP/SMT. The temperature at either location should not exceed 120 °C. The output power of the module should not exceed the rated power for the module (VO, set x IO, max). The SMT05 thermal data presented is based on measurements taken in a wind tunnel. The test setup shown in Figure 7 and EUT need to solder on  $33mm \times 40.38mm(1.300" \times 1.59")$  test pcb. Note that airflow is parallel to the long axis of the module as shown in Fig7.



#### 7.4 Power De-Rating Curves

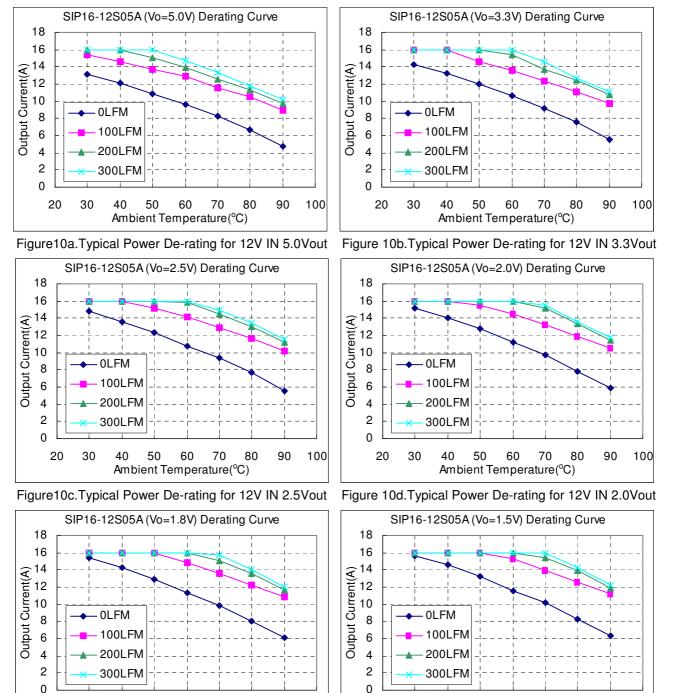
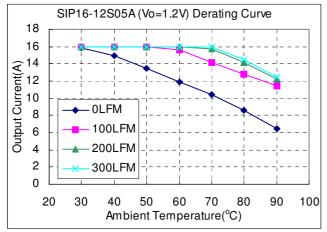


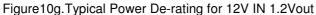
Figure 10e. Typical Power De-rating for 12V IN 1.8Vout Figure 10f. Typical Power De-rating for 12V IN 1.5Vout

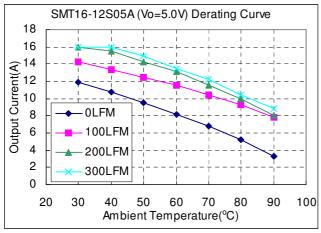
Ambient Temperature(°C)

Ambient Temperature(°C)

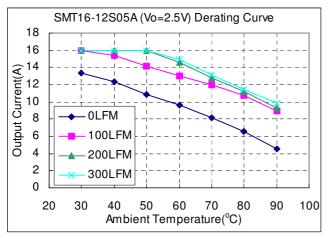


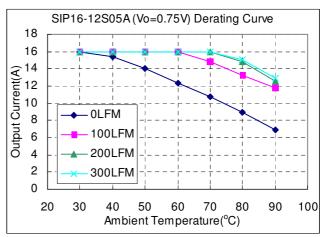




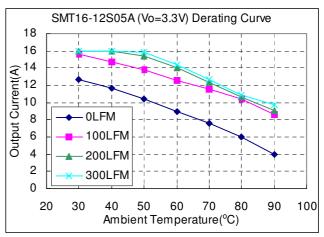


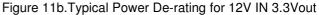


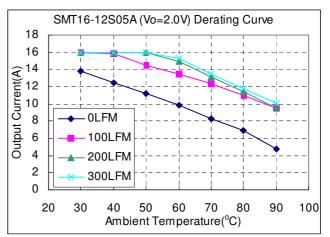






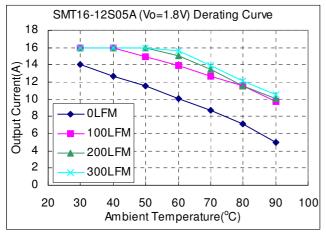


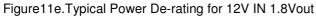


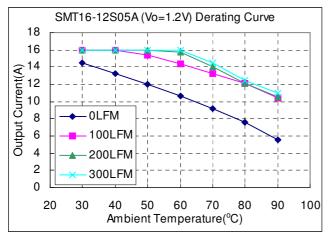




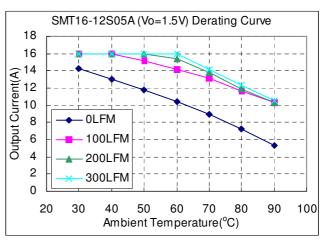














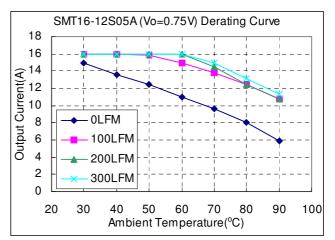
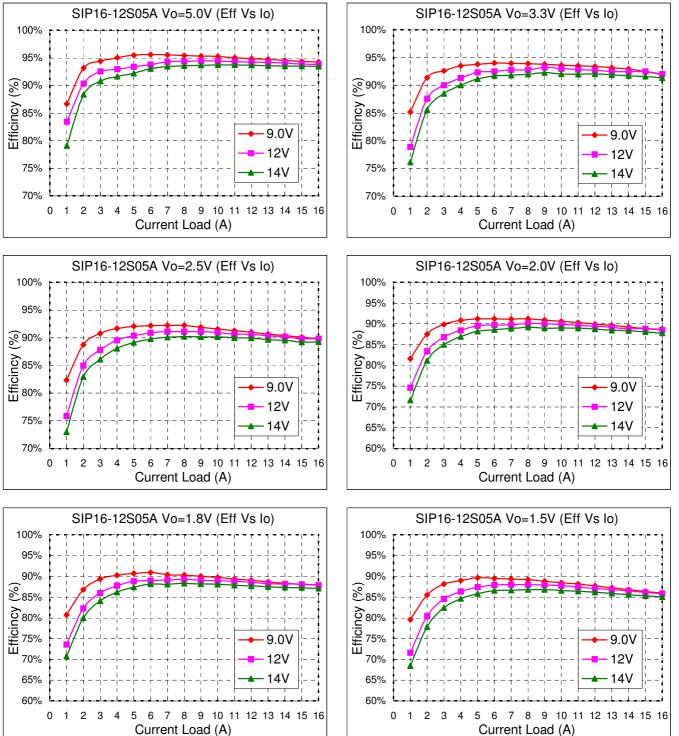


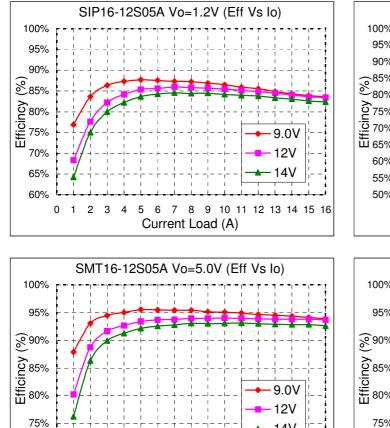
Figure 11f.Typical Power De-rating for 12V IN 1.2Vout Figure 11g.Typical Power De-rating for 12V IN 0.75Vout



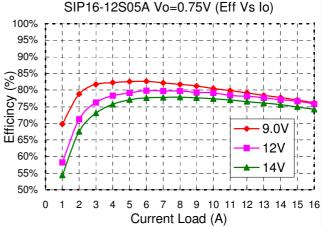
#### 7.5 Efficiency vs Load Curves

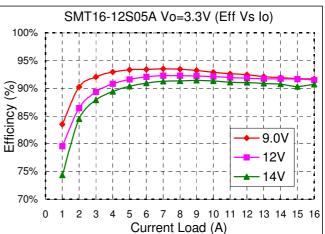


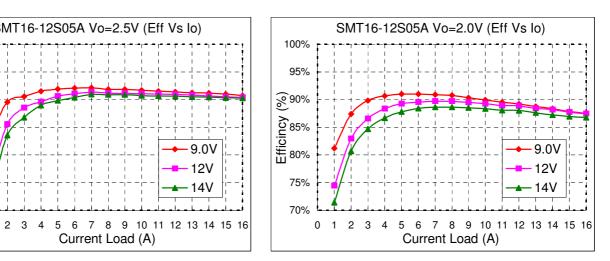


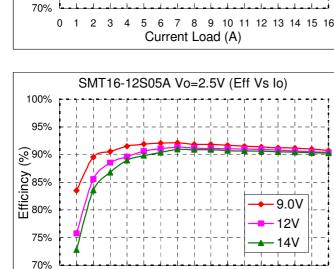


14V







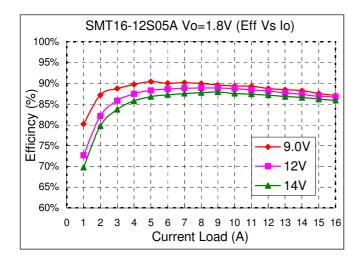


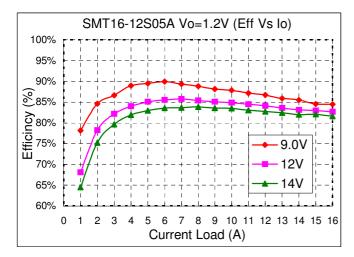
Current Load (A)

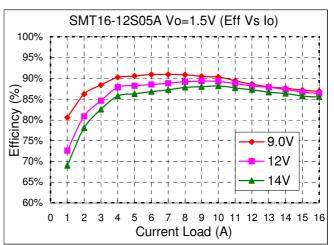
0 1

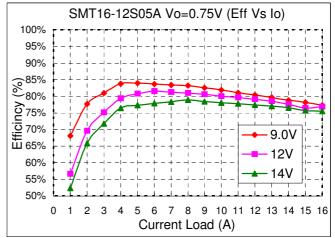


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#### 7.6 Input Capacitance at the Power Module

The SIP/SMT converters must be connected to a low AC source impedance. To avoid problems with loop stability source inductance should be low. Also, the input capacitors should be placed close to the converter input pins to de-couple distribution inductance. However, the external input capacitors are chosen for suitable ripple handling capability. Low ESR polymers are a good choice. They have high capacitance, high ripple rating and low ESR (typical <100mohm). Electrolytic capacitors should be avoided. Circuit as shown in Figure 14 represents typical measurement methods for ripple current. Input reflected-ripple current is measured with a simulated source Inductance of 1uH. Current is measured at the input of the module.

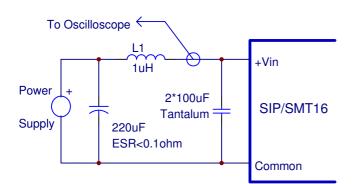


Figure 14. Input Reflected-Ripple Test Setup

#### 7.7 Test Set-Up

The basic test set-up to measure parameters such as efficiency and load regulation is shown in Figure 15. Things to note are that this converter is non-isolated, as such the input and output share a common ground. These grounds should be connected together via low impedance ground plane in the application circuit. When testing a converter on a bench set-up, ensure that -Vin and -Vo are connected together via a low impedance short to ensure proper efficiency and load regulation measurements are being made. When testing the Cincon's SIP/SMT16-12 series under any transient conditions please ensure that the transient response of the source is sufficient to power the equipment under test. We can calculate the

Efficiency

Load regulation and line regulation.

The value of efficiency is defined as:

$$\eta = \frac{Vo \times Io}{Vin \times Iin} \times 100\%$$

Where: Vo is output voltage, lo is output current, Vin is input voltage, lin is input current.

The value of load regulation is defined as:

$$Load.reg = \frac{V_{FL} - V_{NL}}{V_{NL}} \times 100\%$$

The value of line regulation is defined as:

$$Line.reg = \frac{V_{HL} - V_{LL}}{V_{U}} \times 100\%$$

Where:  $V_{HL}$  is the output voltage of maximum input voltage at full load.  $V_{LL}$  is the output voltage of minimum input voltage at full load.

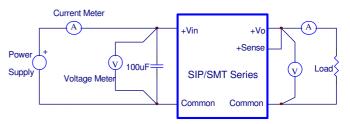


Figure 15. SIP/SMT16 Series Test Setup

#### 7.8 Remote Sense Compensation

Remote Sense regulates the output voltage at the point of load. It minimizes the effects of distribution losses such as drops across the connecting pin and PCB tracks (see Figure 16). Please note however, the maximum drop from the output pin to the point of load should not exceed 500mV for remote compensation to work.

The amount of power delivered by the module is defined as the output voltage multiplied by the output current (VO  $\times$  IO).

When using TRIM UP, the output voltage of the module will increase which, if the same output current is maintained, increases the power output by the module. Make sure that the maximum output power of the module remains at or below the maximum rated power.

When the Remote Sense feature is not being used, leave sense pin disconnected.

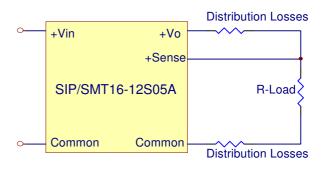


Figure 16. Circuit Configuration for Remote Sense Operation



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#### 7.8 SIP/SMT16-12 Series Output Voltage Adustment.

The output Voltage of the SIP/SMT16-12S05A can be adjusted in the range 0.75V to 5.0V by connecting a single resistor on the motherboard (shown as Rtrim) in Figure 16. When Trim resistor is not connected the output voltage defaults to 0.75V

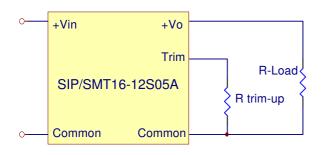


Figure 16. Trim-up Voltage Setup The value of Rtrim-up defined as:

$$Rtrim = (\frac{10500}{Vo - 0.75} - 1000)$$

Where: Rtrim-up is the external resistor in ohm,

Vo is the desired output voltage

To give an example of the above calculation, to set a voltage of 3.3Vdc, Rtrim is given by:

$$Rtrim = (\frac{10500}{3.3 - 0.75} - 1000)$$

Rtrim = 3117 ohm

For various output values various resistors are calculated and provided in Table 3 for convenience.

Vo,set (V)	Rtrim (Kohm)
0.75	Open
1.20	22.33
1.50	13.0
1.80	9.0
2.00	7.4
2.50	5.0
3.30	3.12
5.0	1.47

Table 3 - Trim Resistor Values

#### 7.9 Output Ripple and Noise Measurement

The test set-up for noise and ripple measurements is shown in Figure 17. a coaxial cable with a 50ohm termination was used to prevent impedance mismatch reflections disturbing the noise readings at higher frequencies.

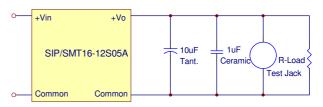


Figure 17. Output Voltage Ripple and Noise Measurement Set-Up

#### 7.10 Output Capacitance

Cincon's SIP/SMT16-12 series converters provide unconditional stability with or without external capacitors. For good transient response low ESR output capacitors should be located close to the point of load. For high current applications point has already been made in layout considerations for low resistance and low inductance tracks. Output capacitors with its associated ESR values have an impact on loop stability and bandwidth. Cincon's converters are designed to work with load capacitance up-to 8,000uF. It is recommended that any additional capacitance, Maximum 8,000uF and low ESR, be connected close to the point of load and outside the remote compensation point.

#### 7.11 SMT Reflow Profile

An example of the SMT reflow profile is given in Figure 18. Equipment used: SMD HOT AIR REFLOW HD-350SAR Alloy: AMQ-M293TA or NC-SMQ92 IND-82088 SN63

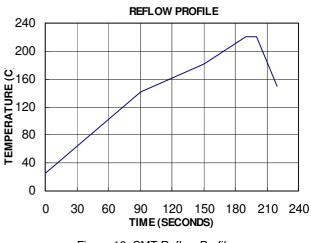


Figure 18 SMT Reflow Profile



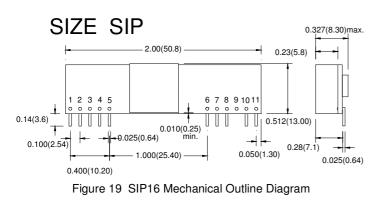
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### 8. Mechanical Outline Diagrams

#### 8.1 SIP/SMT16 Mechanical Outline Diagrams

Dimensions are in millimeters and inches

Tolerance: x.xx  $\pm 0.02$  in. (0.5mm) , x.xxx  $\pm 0.010$  in. (0.25 mm) unless otherwise noted



BOTTOM VIEW OF BOARD

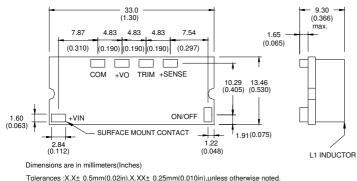
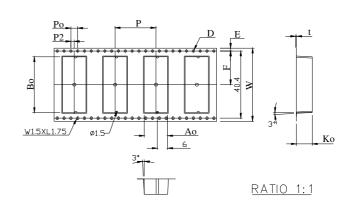


Figure 20 SMT16 Mechanical Outline Diagram

#### 8.2 SMT Tape and Reel Dimensions

The Tape Reel dimensions for the SMT module is shown in Figure 21.



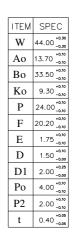


Figure 21 – SMT Tape and Reel Dimensions

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